# Radiometrix

Hartcran House, 231 Kenton Lane, Harrow, HA3 8RP, England Tel: +44 (0) 20 8909 9595, Fax: +44 (0) 20 8909 2233



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# 12.5kHz Channel VHF TX / RX

The ENT1 and ENR1 form a miniature VHF radio transmitter/receiver pair designed for PCB mounting and suitable for extended range wireless data links at speeds up to 3kbps.

Wireless link ranges of 10km+ are achievable with suitable choice of data rate and antennas.



Figure 1: ENT1 & ENR1

Available for operation at 169.44375MHz or 169.40625MHz in Europe, the ENT1 and the ENR1 combine full screening with internal filtering to ensure EMC compliance by minimising spurious radiation and susceptibility. The modules suits one-to-one and multi-node wireless links in applications including automatic meter reading (AMR), car and building security, EPOS and inventory tracking and remote industrial process monitoring.

#### **Features**

- Conforms to EN 300 220-3 (Radio) and EN 301 489-3 (EMC)
- Versions available on 169.44375MHz and 169.40625MHz
- 12.5 kHz channel spacing
- Data rates up to 3kbps
- Usable range over 10km
- Fully screened
- Low power requirements

# **Applications**

- Remote meter reading
- Industrial telemetry and telecommand
- In-building environmental monitoring and control
- High-end security and fire alarms

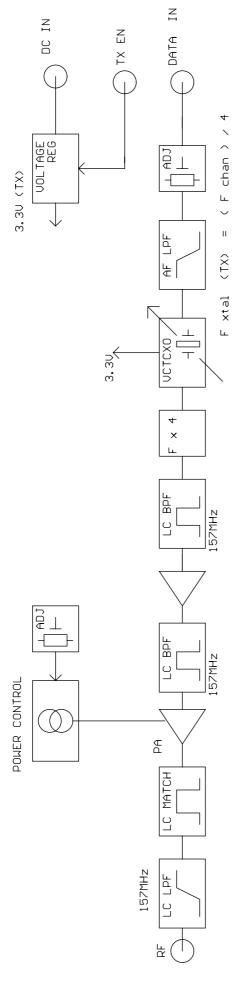
# **Technical Summary**

# ENT1

- Supply range: 5v
- Current consumption: 70mA Max
- Data bit rate: 3kbps max.
- Transmit power: 20dBm (100mW) nominal
- Size: 43 x 14.5 x 5mm

# ENR<sub>1</sub>

- Double conversion FM superhet
- SAW front end filter gives >80dB image rejection
- Operation from 2.7V to 16V @ 13mA typical
- Data bit rate: 5kbps max.
- -115dBm sensitivity @ 1ppm BER
- Size: 48 x 17.5 x 7.2mm



Single channel 100mW UHF transmitter

Figure 2: ENT1 block diagram

Radiometrix Ltd., ENT1/ENR1 data sheet Page 2

# Functional description: ENT1 transmitter

The ENT1 transmitter consists of a frequency modulated Voltage Controlled Temperature Controlled Crystal Oscillator (VCTCXO) feeding a frequency multiplier with two stage amplifier and RF filter. Final Power Amplifier stage is factory pre-set to appropriate band power level. Operation can be controlled by the EN (Enable) line, the transmitter achieving full RF output typically within 7ms of this line being pulled high. The RF output is filtered to ensure compliance with the appropriate radio regulations and fed to the  $50\Omega$  antenna pin.

#### User interface

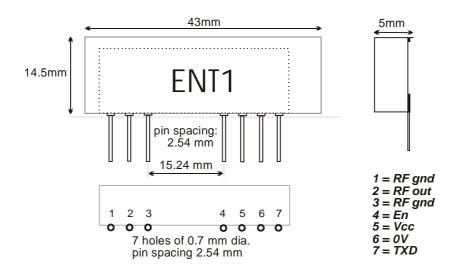


Figure 3: ENT1 pin-out and dimension

# **RF GND** (Pins 1 & 3)

RF ground, internally connected to the module screen and pin 6 (0V). These pins should be directly connected to the RF return path - e.g. coax braid, main PCB ground plane etc.

#### **RF out** (Pin 2)

 $50\Omega$  RF output to antenna. Internally DC-isolated. See antenna section of applications notes for details of suitable antennas / feeds.

#### **En** (Pin 4)

Tx enable.  $\leq$ 0.15V or open-circuit on this pin disables module (current <1mA),  $\geq$ 1.7V enables module. Input impedance 1M $\Omega$  approx. EN pin should not be left floating. Observe slew rate requirements (see applications note).

Vcc (Pin 5)

5V DC +ve supply. Max ripple content 0.1Vp-p. Decoupling is not generally required.

#### **0V** (Pin 6)

DC supply ground. Internally connected to pins 1, 3 and module screen.

#### **TXD** (Pin 7)

DC coupled input for 5V CMOS logic. Rin =  $100k\Omega$ 

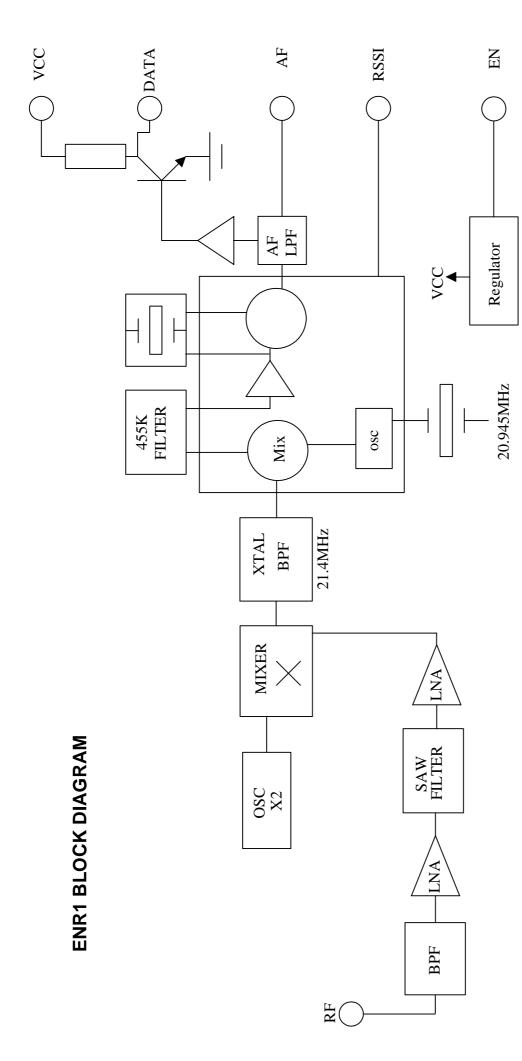


Figure 4: ENR1 block diagram

# Functional description: ENR1 receiver

The ENR1 module is a double conversion NBFM superhet receiver capable of handling data rates of up to 5kbps. It will operate from a supply of 2.7V to 16V and draws 13mA when receiving. A signal strength (RSSI) output with greater than 45dB of range is provided. The SIL style ENR1 measures  $48 \times 17.5 \times 7.2 \text{ mm}$  excluding the pin

# User interface

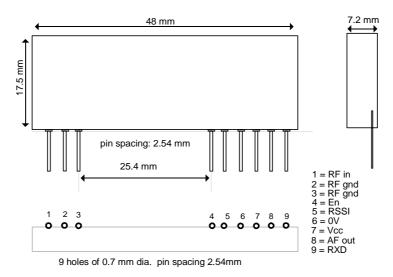


Figure 5: ENR1 pin-out and dimension

# **RF IN** (pin 1)

 $50\Omega$  input from the antenna, DC isolated.

# **RF GND** (pin 2/3)

RF ground pin, internally connected to the module screen and pin 4 (0V). This pin should be connected to the RF return path (coax braid, main PCB ground plane etc.)

### **En** (pin 4)

Rx enable. <0.15V shuts down module (current <1uA). >1.7V enables the receiver. Impedance ~1Mohm. Observe sle rate requirements (see apps notes).

# **RSSI** (pin 5)

Received signal strength indicator with 45dB range. See page 8 for typical characteristics.

# **0V** (pin 6)

DC supply ground. Internally connected to pin 2 and module screen.

#### **Vcc** (pin 7)

DC supply. Max ripple content 0.1V<sub>p-p</sub>.

#### $\mathbf{AF}$ out (pin 8)

Buffered and filtered analogue output from the FM demodulator. Standing DC bias 0.75V approx. External load should be >10k $\Omega$  // <100pF.

#### **RXD** (pin 9)

Digital output from the internal data slicer. The data is a squared version of the signal on pin 8 (AF) and is true data, i.e. as fed to the transmitter. Output is "open-collector" format with internal  $10k\Omega$  pull-up to Vcc (pin 7).

# Absolute maximum ratings

Exceeding the values given below may cause permanent damage to the module.

#### ENT1

Operating temperature  $-40^{\circ}\text{C}$  to  $+80^{\circ}\text{C}$ Storage temperature  $-40^{\circ}\text{C}$  to  $+100^{\circ}\text{C}$ 

RF out (pin 2) ±50V @ <10MHz, +20dBm @ >10MHz

All other pins -0.3V to +5.5V

#### ENR1:

 $\begin{array}{lll} \text{Operating temperature} & -20^{\circ}\text{C to } +55^{\circ}\text{C} \\ \text{Storage temperature} & -40^{\circ}\text{C to } +100^{\circ}\text{C} \\ \text{Vcc (pin 7)} & -0.1\text{V to } +16\text{V} \\ \text{RSSI, AF, RXD (pins 5,8,9)} & -0.1\text{V to } +3\text{V} \\ \end{array}$ 

RF IN (pin 1) ±50V DC, +10dBm RF

# Performance specifications: ENT1

 $(Vcc = 5V / temperature = 20 \, ^{\circ}C \, unless \, stated)$ 

	pin	min.	typ.	max.	units	notes
DC supply						
Supply voltage	5 5	4.5	5	5.5	V	
TX Supply current @ 100mW			65mA		mA	
Antenna pin impedance	2		50		$\Omega$	
RF centre frequency (100mW)			169.406250		MHz	6
Channel spacing			12.5		kHz	
Number of channels			1			
RF						
RF power output	2	+19	+20	+21	dBm	1
Spurious emissions	2				dBm	5
Adjacent channel TX power			-40		dBm	
Frequency accuracy		-1.5	0	+1.5	kHz	2
FM deviation (peak)		$\pm 1.4$	±1.5	$\pm 1.6$	kHz	3
Baseband						
Modulation bandwidth @ -3dB		0	-	2	kHz	
TXD input level (logic low)	7		0		V	4
TXD input level (logic high)	7		5		V	4
Dynamic timing						
TX select to full RF			7		ms	

#### Notes:

- 1. Measured into  $50\Omega$  resistive load.
- 2. Total over full supply and temperature range.
- 3. With 0V 5.0V modulation input.
- 4. To achieve specified FM deviation.
- 5. Meets EN300-220
- 6. Available on other frequencies from 120MHz to 180MHz (subject to MOQ and lead time)

# Performance specifications: ENR1 receiver

(Vcc = 5.0V / temperature = 20 % unless stated)

(Vcc = 3.0V / temperature = 20 % ui	pin	min.	typ.	Max.	units	notes
DC supply	<i>y</i>	******	<u> </u>	WWW	diffes _	Hotos
~ , ,	_					
Supply voltage	7	2.7	5.0	16	V.	
Supply current	7	10	13	15	mA	
RF/IF						
RF sensitivity for 12dB (S+N/N)	1, 8	-118	-120	-	dBm	1,2
RF sensitivity for 1ppm BER	1, 9	-	-115	-	dBm	1,2
RSSI range	1, 5	-	45	-	dB	
IF bandwidth		-	7.5	-	kHz	
Image rejection	1	70	89	-	dB	
LO leakage, conducted	1	-54	-65		dBm	
Adjacent channel rejection			<-70		dP	
Blocking			>85		dE	
Baseband					.\\'	
Baseband bandwidth @ -3dB	8	0		3	kНz	
AF level	8	300	400	450	riV <sub>P-P</sub>	1
DC offset on AF out	8	0.5	0.75	1.25	V	1
Distortion on recovered AF	8	0.5	0 / 3	5	\ \frac{\dot{\dot{V}}}{\%}	1,2
Load capacitance, AFout/RXD	3,8			100	pF	1,2
Load capacitance, Arout/ICAD	3,0	ĪΛ'		100	hī.	
DYNAMIC TIMING		$\langle       \rangle$				
		A				
Power up with signal present	111					
Power up to valid RSSI	2,5	<b>\</b> -	6.5	7.5	ms	
Power up to stable data	39	-	10	13	ms	
Signal applied with supriy on			0.4			
RSSI response time (rise/iall)	1,5	-	0.1	-	ms	
Signal to stable daya	1,9	-	3.5	-	ms	

# Notes:

- 1. For received signal with  $\pm 1.5 kHz$  FM deviation. 2. Typical figures are for signal at centre frequency

# **Applications information**

# Power supply requirements

The ENT1 transmitter requires a regulated 5V supply, but the ENR1 receiver incorporates a built-in regulator which delivers a constant 2.8V to the module circuitry when the external supply voltage is 2.85V or greater, with 40dB or more of supply ripple rejection. This ensures constant performance up to the maximum permitted supply rail and removes the need for external supply decoupling except in cases where the supply rail is extremely poor (ripple/noise content >0.1Vp-p).

Note, however, that for supply voltages lower than 2.85V the regulator is effectively inoperative and supply ripple rejection is considerably reduced. Under these conditions the ripple/noise on the supply rail should be below 10mVp-p to avoid problems. If the quality of the supply is in doubt, it is recommended that a  $10\mu F$  low-ESR tantalum or similar capacitor be added between the module supply pin (Vcc) and ground, together with a  $10\Omega$  series feed resistor between the Vcc pin and the supply rail.

# TX modulation requirements

The module is factory-set to produce the specified FM deviation with a TXD input to pin 14 of 5V amplitude, i.e. 0V "low", 5V "high

If the data input level is greater than 5V, a resistor must be added in series with the TXD input to limit the modulating input voltage to a maximum of 5V on pin 7. TXD input resistance is  $100k\Omega$  to ground.

# Received Signal Strength Indicator (RSSI)

The module incorporates a wide range RSSI which measures the strength of an incoming signal over a range of approximately 45dB. This allows assessment of link quality and available margin and is useful when performing range tests.

The output on pin 5 of the module has a standing DC bias in the region of 0.6V with no signal, rising to around 1.75V at maximum indication. The RSSI output source impedance is high  $(\sim 100 \mathrm{k}\Omega)$  and external loading should therefore be kept to a minimum.

Typical RSSI characteristic is as shown below:

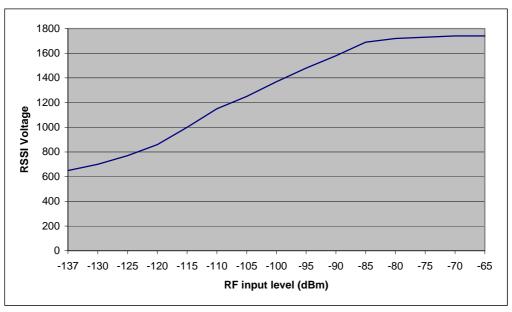


Figure 6: ENR1 RSSI response curve

# **Expected range**

Predicting the range obtainable in any given situation is notoriously difficult since there are many factors involved. The main ones to consider are as follows:

- Type and location of antennas in use
- Type of terrain and degree of obstruction of the link path
- Sources of interference affecting the receiver
- "Dead" spots caused by signal reflections from nearby conductive objects
- · Data rate and degree of filtering employed

The following are typical examples – but range tests should always be performed before assuming that a particular range can be achieved in a given situation:

Data rate	Tx antenna	Rx antenna	Environment	Range
1.2kbps	half-wave	half-wave	rural/open	10-15km

The ENT1's TXD input is normally driven directly by logic levels but will also accept analogue drive (e.g. 2-tone signalling). In this case it is recommended that TXD (pin 14) be DC-biased to 2.5V approx. with the modulation ac-coupled and limited to a maximum of  $5V_{p-p}$  to minimise distortion over the link.

Although the modulation bandwidth of the ENT1 extends down to DC it is not advisable to use data containing a DC component. This is because frequency errors and drifts between the transmitter and receiver occur in normal operation, resulting in DC offset errors on the receiver's audio output.

#### **Antennas**

The choice and positioning of transmitter and receiver antennas is of the utmost importance and is the single most significant factor in determining system range. The following notes are intended to assist the user in choosing the most effective antenna type for any given application.

# Integral antennas

These are relatively inefficient compared to the larger externally-mounted types and hence tend to be effective only over limited ranges. They do however result in physically compact equipment and for this reason are often preferred for portable applications. Particular care is required with this type of antenna to achieve optimum results and the following should be taken into account:

- 1. Nearby conducting objects such as a PCB or battery can cause detuning or screening of the antenna which severely reduces efficiency. Ideally the antenna should stick out from the top of the product and be entirely in the clear, however this is often not desirable for practical/ergonomic reasons and a compromise may need to be reached. If an internal antenna must be used try to keep it away from other metal components and pay particular attention to the "hot" end (i.e. the far end) as this is generally the most susceptible to detuning. The space around the antenna is as important as the antenna itself.
- 2. Microprocessors and microcontrollers tend to radiate significant amounts of radio frequency hash which can cause desensitisation of the receiver if its antenna is in close proximity. The problem becomes worse as logic speeds increase, because fast logic edges generate harmonics across the VHF range which are then radiated effectively by the PCB tracking. In extreme cases system range may be reduced by a factor of 5 or more. To minimise any adverse effects situate antenna and module as far as possible from any such circuitry and keep PCB track lengths to the minimum possible. A ground plane can be highly effective in cutting radiated interference and its use is strongly recommended.

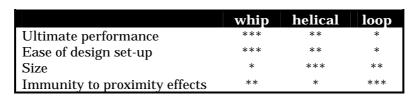
#### The following types of integral antenna are in common use:

**Quarter-wave whip.** This consists simply of a piece of wire or rod connected to the module at one end. At 169MHz the total length should be 421mm from module pin to antenna tip including any interconnecting wire or tracking. Because of the length of this antenna it is almost always external to the product casing.

*Helical.* This is a more compact but slightly less effective antenna formed from a coil of wire. It is very efficient for its size, but because of its high Q it suffers badly from detuning caused by proximity to nearby conductive objects and needs to be carefully trimmed for best performance in a given situation. The size shown is about the maximum commonly used at 169MHz and appropriate scaling of length, diameter and number of turns can make individual designs much smaller.

**Loop.** A loop of PCB track having an inside area as large as possible (minimum about 5cm²), tuned and matched with 2 capacitors. Loops are relatively inefficient but have good immunity to proximity detuning, so may be preferred in shorter range applications where high component packing density is necessary.

Integral antenna summary:



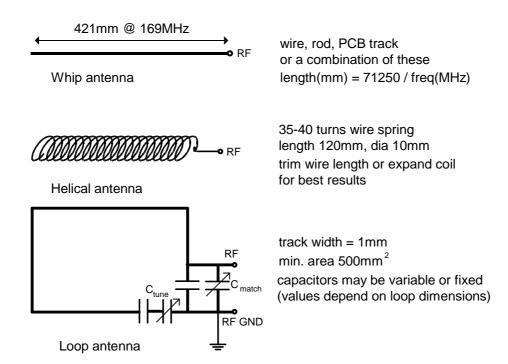


Figure 7: integral antenna configurations

# External antennas

These have several advantages if portability is not an issue, and are essential for long range links. External antennas can be optimised for individual circumstances and may be mounted in relatively good RF locations away from sources of interference, being connected to the equipment by coax feeder.

*Helical.* Of similar dimensions and performance to the integral type mentioned above, commercially-available helical antennas normally have the coil element protected by a plastic moulding or sleeve and incorporate a coax connector at one end (usually a straight or right-angle BNC type). These are compact and simple to use as they come pre-tuned for a given application, but are relatively inefficient and are best suited to shorter ranges.

**Quarter-wave whip.** Again similar to the integral type, the element usually consists of a stainless steel rod or a wire contained within a semi-flexible moulded plastic jacket. Various mounting options are available, from a simple BNC connector to wall brackets, through-panel fixings and magnetic mounts for temporary attachment to steel surfaces.

A significant improvement in performance is obtainable if the whip is used in conjunction with a metal ground plane. For best results this should extend all round the base of the whip out to a radius of 300mm or more (under these conditions performance approaches that of a half-wave dipole) but even relatively small metal areas will produce a worthwhile improvement over the whip alone. The ground plane should be electrically connected to the coax outer at the base of the whip. Magnetic mounts are slightly different in that they rely on capacitance between the mount and the metal surface to achieve the same result.

A ground plane can also be simulated by using 3 or 4 quarter-wave radials equally spaced around the base of the whip, connected at their inner ends to the outer of the coax feed. A better match to a  $50\Omega$  coax feed can be achieved if the elements are angled downwards at approximately 30-40° to the horizontal.

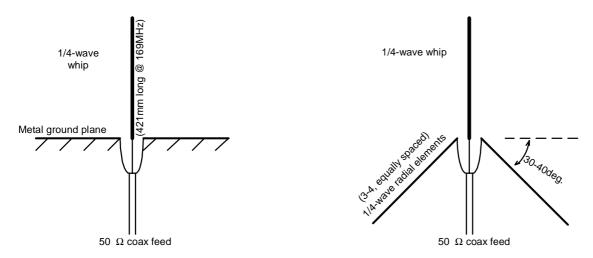


Fig.8: Quarter wave antenna / ground plane configurations

*Half-wave.* There are two main variants of this antenna, both of which are very effective and are recommended where long range and all-round coverage are required:

- The half-wave dipole consists of two quarter-wave whips mounted in line vertically and fed in the
  centre with coaxial cable. The bottom whip takes the place of the ground plane described previously.
  A variant is available using a helical instead of a whip for the lower element, giving similar
  performance with reduced overall length. This antenna is suitable for mounting on walls etc. but for
  best results should be kept well clear of surrounding conductive objects and structures (ideally >1m
  separation).
- 2. The end-fed half wave is the same length as the dipole but consists of a single rod or whip fed at the bottom via a matching network. Mounting options are similar to those for the quarter-wave whip. A ground plane is sometimes used but is not essential. The end-fed arrangement is often preferred over the centre-fed dipole because it is easier to mount in the clear and above surrounding obstructions.

**Yagi.** This antenna consists of two or more elements mounted parallel to each other on a central boom. It is directional and exhibits gain but tends to be large and unwieldy – for these reasons the yagi is the ideal choice for links over fixed paths where maximum range is desired.

# Module mounting considerations

The modules may be mounted vertically or bent horizontal to the motherboard.

Good RF layout practice should be observed. If the connection between module and antenna is more than about 20mm long use  $50\Omega$  microstrip line or coax or a combination of both. It is desirable (but not essential) to fill all unused PCB area around the module with ground plane.

If the connection between module and antenna does not form part of the antenna itself, it should be made using  $50\Omega$  microstrip line or coax or a combination of both. It is desirable (but not essential) to fill all unused PCB area around the module with ground plane.

The module may be potted if required in a viscous compound which cannot enter the screen can.

Warning: DO NOT wash the module. It is not hermetically sealed.

# Variants and ordering information

The ENT1 and ENR1 are manufactured on the 169.44375MHz and 169.40625MHz European frequency allocation as standard.

ENT1-169.40625-3 ENR1-169.40625-3

ENT1-169.44375-3 ENR1-169.44375-3

RF output of ENT1 can also be factory set from +5dBm (3mW) to +20dBm (100mW) depending on minimum order quantity.

**Note**: Other variants of ENT1 and ENR1 can be supplied to individual customer requirements at frequencies from 120MHz to 180MHz and/or optimised for specific data speeds and formats. However these are subject to minimum order quantity (MOQ) and long lead time. Please consult the Sales Department for further information.

# **Notes**

# **Radiometrix Ltd**

Hartcran House 231 Kenton Lane Harrow, Middlesex HA3 8RP ENGLAND

Tel: +44 (0) 20 8909 9595 Fax: +44 (0) 20 8909 2233 sales@radiometrix.com www.radiometrix.com

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The Intrastat commodity code for all our wireless modules is: 8542 6000

# **R&TTE Directive**

After 7 April 2001 the manufacturer can only place finished product on the market under the provisions of the R&TTE Directive. Equipment within the scope of the R&TTE Directive may demonstrate compliance to the essential requirements specified in Article 3 of the Directive, as appropriate to the particular equipment. Further details are available on The Office of Communications (Ofcom) web site:

www.ero.dk

http://www.ofcom.org.uk/radiocomms/ifi/

Information Requests
Ofcom
Riverside House
2a Southwark Bridge Road
London SE1 9HA
Tel: +44 (0)845 456 3000 or 020 7981 3040

Fax: +44 (0)20 7783 4033

information.requests@ofcom.org.uk

European Radiocommunications Office (ERO) Peblingehus Nansensgade 19 DK 1366 Copenhagen Tel. +45 33896300 Fax +45 33896330 ero@ero.dk